MSK

100 VOLT 20 AMP H-BRIDGE PWM MOTOR DRIVER/AMPLIFIER

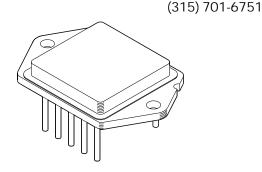
4206

M.S.KENNEDY CORP.

4707 Dey Road Liverpool, N.Y. 13088

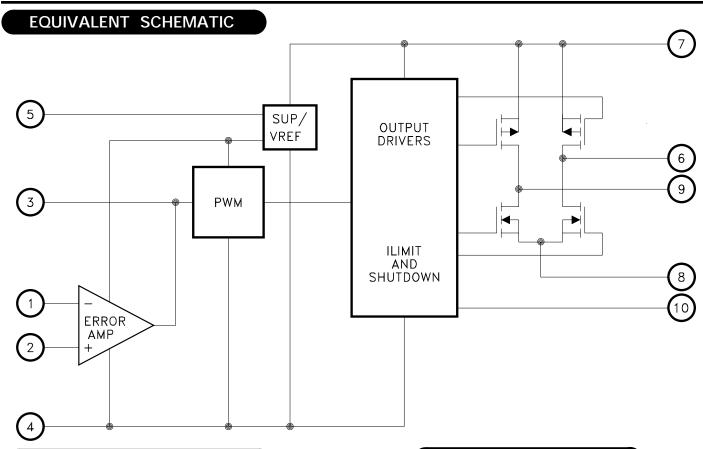
FEATURES:

- Replaces APEX SA01 PWM Amplifier
- 100 Volt, 20 Amp Capability
- · Self-Contained Smart Lowside/Highside Drive Circuitry
- Internal PWM Generation, Shoot-through Protection
- Isolated Case Allows Direct Heatsinking
- Available Fully Screened To MIL-PRF-38534 Class H
- · Contact MSK for MIL-PRF-38534 Qualification Status



DESCRIPTION:

The MSK 4206 is a complete H-Bridge microcircuit intended for use in DC brushed motor control applications or Class D switchmode amplification in severe operating conditions. All of the control circuitry, including an error amplifier, is internal to the device. A precision reference output is provided for use in offsetting the error amplifier. Fixed highside current limit is provided internally. Lowside current limit is user controlled with the use of an external resistor network. The design is packaged in a 10 pin bolt down power package.



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TYPICAL APPLICATIONS

- MOTOR CONTROL
- REACTIVE LOADS
- PIEZO ELEMENTS

PIN-OUT INFORMATION

1 - IN 6 A OUT 2 + IN 7 + VS 3 EA OUT 8 ISENSE 4 GND 9 B OUT

5 REF 10 SHDN/FILTER

ABSOLUTE MAXIMUM RATINGS

Tst	Storage Temperature Range65°C to +150°C
TLD	Lead Temperature Range
	(10 Seconds)
Tc	Case Operating Temperature Range
	MSK4206H/E
	MSK4206 -40° C to $+85^{\circ}$ C

TJ Junction Temperature + 150°C

(Output FETS @ 25°C) 1.0°C/W

θJc Thermal Resistance

ELECTRICAL SPECIFICATIONS

				(
Parameter	Test Conditions	Group A Subgroup	MSK 4206 H/E Min. Typ. Max.		MSK 4206 Min. Typ. Max.			Units	
OUTPUT CHARACTERISTICS									
		1	-	1.60	2.0	-	1.60	2.2	V
VDS(ON) Highside 6	ID = 10A	2	-	2.4	3.0	-	-	-	V
		3	-	1.1	1.5	-	-	-	V
	ID=10A	1	-	0.75	1.0	-	0.75	1.2	V
VDS(ON) Lowside 6		2	-	1.1	1.5	-	-	-	V
		3	-	0.45	0.70	-	-	-	V
	ID = 20A	1	-	3.5	-	-	3.5	-	V
VDS(ON) Highside ② ⑥	ID = 15A	2	-	3.7	-	-	-	-	V
	ID = 20A	3	-	2.1	-	-	-	-	V
		1	-	1.5	-	-	1.5	-	V
VDS(ON) Lowside ② ⑥	ID = 20A	2	-	2.4	-	-	-	-	V
		3	-	0.90	-	-	-	-	V
	D OLLANDIEL	1	-	110	125	-	110	125	mΩ
De L'ANDREST (A. C.)	P-CHANNEL	2	-	168	250	-	_	-	mΩ
RDS(ON) each MOSFET ②⑥⑦ ID=10A		1	-	22	52	-	22	52	mΩ
15 - 10/4	N-CHANNEL	2	-	78	95	-	_	-	mΩ
Leakage Current Highside		1	-	0.02	25	-	0.02	25	uA
Loakago Carront Inghistae	+ Vs = 80V	2	-	1.0	250	-	-	-	uA
		1	-	0.02	25	-	0.02	25	ųΑ
Leakage Current Lowside A/B out=80V		2	-	1.0	250	-	-	-	ųΑ
OUTPUT FREQUENCY and DUTY CYCLE									
Switching Frequency			35	42	49	32	42	52	KHz
EA OUT = 5V Duty Cycle EA OUT = 8V, Output A = 100%, Output B = 0%		4	40	50	60	40	50	60	%
		7	-	Verify	-	_	Verify	-	P/F
EA OUT = 2V, Output A = 0%, Output B = 100%			-	Verify	-	-	Verify	-	P/F
SUPPLY CURRENT CHARACTERISTICS				· · ·			,		
+ Vs (2)		1,2,3	16	50	100	16	50	100	V
+ Vs Current IREF = 5mA, RL = 10K	Ω SHDN/FILTER = < 180mV	1	-	60	93	-	60	93	mA
+ Vs Current IREF = 5mA, RL = 10K	Ω SHDN/FILTER = > 180mV	1	-	20	25	-	-	25	mA
REFERENÇE									
		1	7.46	7.50	7.54	7.42	7.50	7.58	V
Output Voltage	I0u⊤=5mA	2,3	7.42	-	7.58	-	-	-	V
Load Regulation		-	-	20	-	-	20	-	PPM/m/
Line Regulation			-	1	-	-	1	-	PPM/V
Error Amp							-		
•			-10	-	+10	-15	-	+ 15	mV
Offset Voltage		2,3	-15	_	+ 15	-	-	-	mV
			-	0.5	5	-	0.5	8	uА
Input Bias Current		1 2,3	-	0.7	13	-	-	-	uА
			-	0.01	1	-	0.01	2	uA
Offset Current			-	0.01	8	-	-	-	uA
Common Mode Voltage range ②			2	-	8	2	0.01	8	V
Common Mode Rejection, DC ②			75	-	-	75		-	dB
Slew Rate			-	15	-	-	15	_	V/uS
Open loop gain ②			75		-	75	-	_	dB
Gain Bandwidth Product			-	2	_	-		_	MHz
SHDN/FILTER							2		2
Trip Point			180	200	220	170	200	230	mV
•			-			-			uA
Input Current			-	0.8	10	-	0.8	10	u/

NOTES:

- 1 + Vs = 28V, I SENSE = Ground, SHDN/FILTER = 0V unless otherwise specified.
- ② Guaranteed by design but not tested. Typical parameters are representative of actual device performance but are for reference only.
- 3 Class H devices are 100% tested to subgroups 1,2,3,4, and 7. Subgroup 5 and 6 testing available upon request.
- $\overline{4}$ Subgroup 1,4,7 $TA = TC = +25^{\circ}C$
 - 2.5 TA=TC= + 125° C
 - 3,6 $TA = TC = -55^{\circ}C$
- (5) Industrial grade and "E" suffix devices shall be 100% tested at 25°C only.
- (6) Tested with a low duty cycle pulse to minimize junction heating.
- The typical internal reading is for the die only. This should be used for thermal calculations only.
- (8) Continuous operation at or above the maximum ratings may adversely effect the device performance and/or life cycle.

APPLICATION NOTES

MSK 4206 PIN DESCRIPTION

- + **Vs** Is the voltage supply for powering internal logic and drivers as well as H-bridge supply pin. The MOSFETS obtain the drive current from this supply pin. The MOSFETs are rated at 100 volts. Proper bypassing to GND with sufficient capacitance to suppress any voltage transients, and ensure removal of any drooping during switching, should be done as close to the pin on the hybrid as possible.
- **AOUT** Is the output pin for one half of the bridge. As EA OUT goes more positive the duty cycles increases at this output.
- **BOUT** Is the output pin for one half of the bridge. As EA OUT goes more positive the duty cycles decreases at this output.
- **ISENSE** Is the connection for the bottom of the half bridge. This can have a sense resistor connection to the \pm Vs return ground for current limit sensing, or can be connected directly to ground. The maximum voltage on this pin is \pm 5V with respect to GND.
- **GND** Is the return connection for + Vs.
- + IN Is the non-inverting input to the internal error amplifier. This pin requires a level between 2 and 8V for proper operation.
- -IN Is the inverting input to the internal error amplifier.
- **EA OUT** Is the output of the internal error amplifier. The error amplifier output is connected internally to the PWM. An error amp output voltage higher than 5V will produce greater than 50% duty cycle pulses out of A OUT. An error amp voltage lower than 5V will produce greater than 50% duty cycle pulses out of B OUT.
- REF Is the output of the internal precision reference. The internal reference will provide 7.5V with load current up to 5mA.

SHDN/FILTER - Is the connection for disabling all 4 output switches. SHUTDOWN high overrides all other inputs. When taken low, everything functions normally. This pin should be grounded if not used.

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APPLICATION NOTES CONT'D

CURRENT LIMIT

The current limit function sets the peak limit on current flow in pin 8. This circuit can trip at anytime during the conduction period and will hold the output transistors off for the remainder of that conduction period. It is recommended that RLIMIT resistor be non-inductive. Load current flows in the ISENSE pin. To avoid errors due to lead length connect the SHDN/FILTER pin directly to the RLIMIT resistor (through the filter network and shutdown divider resistor) and connect the RLIMIT resistor directly to the + VS return. Switching noise spikes will invariably be found at the ISENSE pin. The noise spikes could trip the current limit threshold which is only 200mV. RFILTER and CFILTER should be adjusted so as to reduce the switching noise well below 200mV to prevent false current limiting. The sum of the DC level plus the noise peak will determine the current limiting value. As in most switching circuits it may be difficult to determine the true noise amplitude without careful attention to grounding of the oscilloscope probe. Use the shortest possible ground lead for the probe and connect exactly at the GND pin of the amplifier. Suggested starting values are CFILTER= 0.01 μ F, RFILTER= 5K. The required value of RLIMIT in voltage mode may be calculated by:

RLIMIT = 0.2V/ILIMIT where RLIMIT is the required resistor value and ILIMIT is the maximum desired current. If RSHDN is used it will further divide down the sense voltage. The shutdown divider network will also have an effect on the filtering circuit.

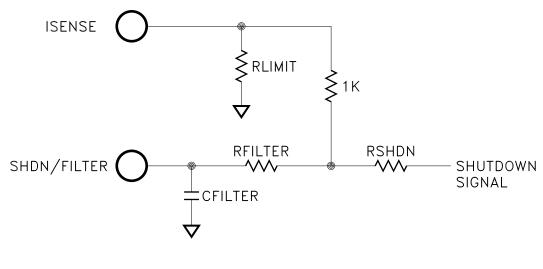


FIGURE 1

ERROR AMPLIFIER

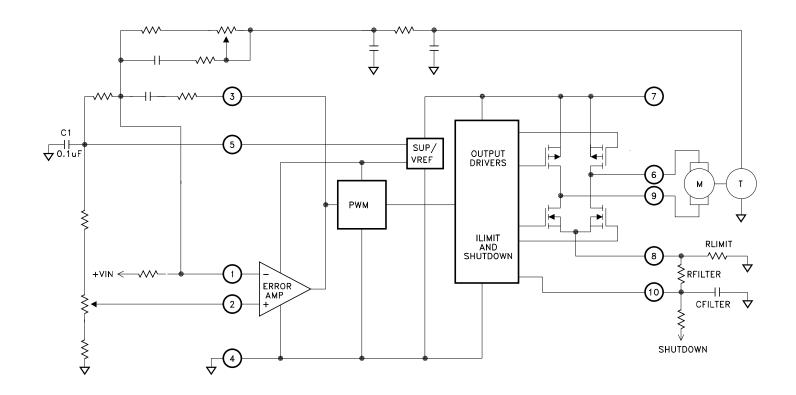
The internal amplifier can be used to integrate the difference between command signals and feedback signals. For highest loop accurancy it is best to configure the error amplifier as an integrator. Poles and zeros must be selected to provide circuit stability while maximizing system respones. Specific component values will vary from system to system. The 7.5V reference output can be used to bias the non-inverting input of the error amplifier to the middle of its 2 to 8V common mode voltage range. (See typical application circuit).

PROTECTION CIRCUITS

In addition to the externally programmable current limit there is also a fixed internal current limit which senses only the highside current. The DC trip level is approximately 35A. Should either of the outputs be shorted to ground the highside current limit will latch off the output transistors. It will be necessary to remove the fault condition and recycle power to + VS to restart the circuit.

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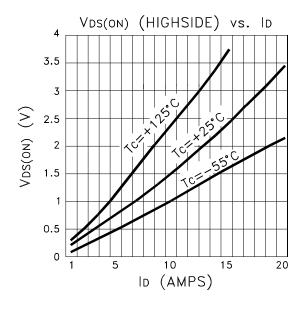


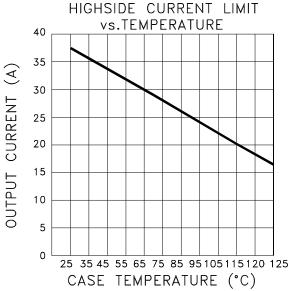
MOTOR DRIVER WITH TACH FEEDBACK

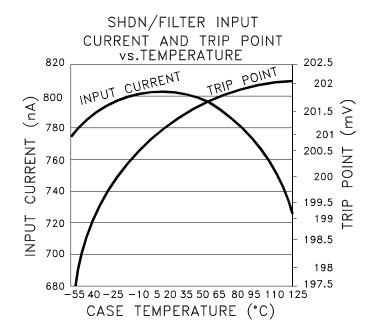
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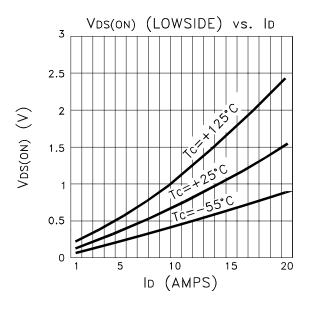
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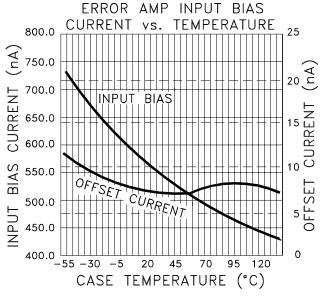
PERFORMANCE CURVES

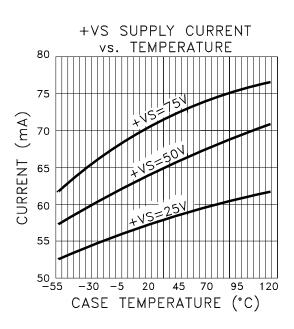








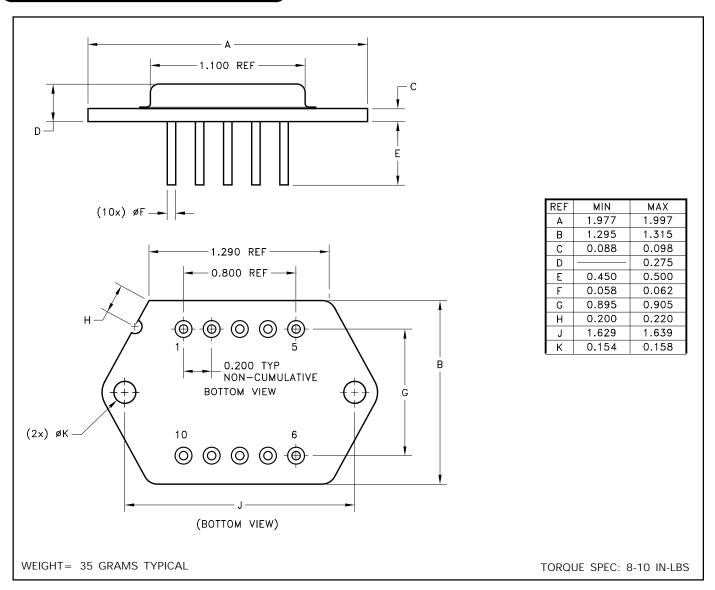




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MECHANICAL SPECIFICATIONS



ORDERING INFORMATION



The above example is a Military grade class H hybrid.

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